

ELECTROMAGNETISM : Course synopsis

Introduction

Brief review of vector knowledge required. Divergence, gradient, curl. Divergence theorem and Stokes theorem + vector identities needed: $\nabla \times \nabla \times \underline{A} = \nabla(\nabla \cdot \underline{A}) - \nabla^2 \underline{A}$, $\nabla \cdot (\underline{A} \times \underline{B}) = \underline{B} \cdot (\nabla \times \underline{A}) - \underline{A} \cdot (\nabla \times \underline{B})$, $\nabla \cdot (\nabla \times \underline{A}) = 0$ always, $\nabla \times (\nabla A) = 0$ always.

Phasor representation. Time-averaged power and related quantities in terms of phasor representation, e.g. $\langle P \rangle = \frac{1}{2} \text{Real} [V(t) \cdot I^*(t)]$.

Reminder of standard expressions for electric and magnetic (Biot-Savart) fields and forces. Lorentz force expression $\underline{F} = q(\underline{E} + \underline{v} \times \underline{B})$

Formulation of differential equation form of Maxwell's Equations:

$$1) \nabla \cdot \underline{E} = \frac{\rho}{\epsilon_0} \quad \text{from Gauss' Law}$$

$$2) \nabla \times \underline{E} = -\frac{\partial \underline{B}}{\partial t} \quad \text{from Faraday's Law of Induction}$$

$$3) \nabla \cdot \underline{B} = 0 \quad \text{no magnetic monopoles}$$

$$4) \nabla \times \underline{B} = \mu_0 \underline{J} + \mu_0 \epsilon_0 \frac{\partial \underline{E}}{\partial t} \quad \text{beginning with Ampere's circuital Law}$$

Reduction to two separate pairs describing electrostatics and magnetostatics when there is no time dependence. Brief references to Laplace's eqn., Poissons eqn. Mention of vector potential defined via $\underline{B} = \nabla \times \underline{A}$.

The Continuity Equation:

$$\nabla \cdot \underline{J} = -\frac{\partial \rho}{\partial t}$$

and charge conservation.

Propagation of EM waves in a vacuum

3-D generalisation of wave eqn. as appropriate for EM waves in free space with $c = 1/(\mu_0 \epsilon_0)^{1/2}$:

$$\nabla^2 \underline{E} = \mu_0 \epsilon_0 \frac{\partial^2 \underline{E}}{\partial t^2}, \quad \nabla^2 \underline{B} = \mu_0 \epsilon_0 \frac{\partial^2 \underline{B}}{\partial t^2}$$

Plane-wave solutions with $\underline{E}, \underline{B} \propto e^{i(\omega t - \underline{k} \cdot \underline{r})}$, $k^2 = \omega^2 \mu_0 \epsilon_0$. Phase and group velocities. Mutual orthogonality of \underline{E} , \underline{B} and \underline{k} with $\underline{E} \times \underline{B}$ in direction of propagation (\underline{k}) and $B = E/c$.

Media

Dielectric materials. Dipole moment \underline{p} , and polarisation \underline{P} . Derivation of expressions for surface charge density $\sigma_b = \underline{P} \cdot \hat{\underline{n}}$, volume charge density $\rho_b = -\nabla \cdot \underline{P}$ and polarisation current density $\underline{J}_b = \partial \underline{P} / \partial t$. The electric displacement, \underline{D} . Electric susceptibility, χ_e , with $\underline{P} = \epsilon_0 \chi_e \underline{E}$ in linear, isotropic, homogeneous (LIH) media.

Magnetic materials. Magnetic dipole moment \underline{m} and magnetisation \underline{M} . Derivation of expressions for volume current density $\underline{J}_M = \nabla \times \underline{M}$ and surface current density $\underline{\lambda}_M = \underline{M} \times \hat{\underline{n}}$. The magnetic field intensity, \underline{H} . Magnetic susceptibility, χ_m , with $\underline{M} = \chi_m \underline{H}$ in LIH media.

Conductivity and decay of excess charge density within an LIH conductor. Relaxation time $\tau = \epsilon / \sigma$.

Maxwell's equations in media

Reformulation and solutions of Maxwell's eqns. in a general LIH medium:

$$1) \nabla \cdot \underline{E} = \frac{\rho_{free}}{\epsilon} \approx 0$$

$$2) \nabla \times \underline{E} = -\frac{\partial \underline{B}}{\partial t}$$

$$3) \nabla \cdot \underline{B} = 0$$

$$4) \nabla \times \underline{B} = \mu \underline{J}_{free} + \mu \epsilon \frac{\partial \underline{E}}{\partial t} = \mu \sigma \underline{E} + \mu \epsilon \frac{\partial \underline{E}}{\partial t}$$

Generalised form of wave equations:

$$\nabla^2 \begin{Bmatrix} \underline{E} \\ \underline{B} \\ \underline{H} \end{Bmatrix} - \mu \epsilon \frac{\partial^2}{\partial t^2} \begin{Bmatrix} \underline{E} \\ \underline{B} \\ \underline{H} \end{Bmatrix} - \mu \sigma \frac{\partial}{\partial t} \begin{Bmatrix} \underline{E} \\ \underline{B} \\ \underline{H} \end{Bmatrix} = 0$$

Specific solutions of $k^2 = \mu \epsilon \omega^2 - i \mu \sigma \omega$ in the poor ($\epsilon \omega \gg \sigma$) and good ($\sigma \gg \epsilon \omega$) conductor limits. Real (n) and complex ($n^* = n_r - i n_i$) refractive indices, in both cases refractive index defined as ck/ω .

For a good conductor $n_r = n_i = c(\mu\sigma/2\omega)^{1/2}$, skin depth $\delta = \left(\frac{2}{\mu\sigma\omega}\right)^{1/2}$,

$v_{ph} = \omega/k_r = c/n_r = (2\omega/\mu\sigma)^{1/2} = v_g/2$. Copper and seawater examples. Transparency of "poor" metals to EM waves in the UV region.

Energy flow and Poynting's vector

Poynting's vector $\underline{S} = \underline{E} \times \underline{H}$. Proof that $P = \int \underline{S} \cdot d\underline{a} = \int (\underline{E} \times \underline{H}) \cdot d\underline{a}$ leads to consistent statement of conservation of energy. Current-carrying wire and simple plane-wave examples. Time-averaged Poynting vector,

$$\underline{S}_{ave} = \frac{1}{2} \text{Re}[\underline{E} \times \underline{H}^*] \text{ when phasor representation employed. } S_{ave} = \frac{E_o H_o}{2} = \frac{E_o^2}{2\mu_o c}$$

for plane-waves in free space. Impedance of free space $[=\mu_o c = (\mu_o/\epsilon_o)^{1/2}]$ and LIH media $[=(\mu/\epsilon)^{1/2}]$. Ratio of electric/magnetic energy densities for non-conducting LIH media (=1) and good conductors ($=\omega\epsilon/\sigma$).

Reflection/refraction at the interface between two media

Derivation of boundary conditions - continuity of tangential components of \underline{E} and \underline{H} at the interface.

Use of boundary condition on \underline{E} to show constancy of frequency for the incident, reflected and transmitted (plane) waves. Proof that $\theta_I = \theta_R$ and Snell's Law result (independent of polarisation). Total internal reflection (TIR), critical angle and existence of evanescent wave with associated decay length. Frustrated TIR.

Dielectric/metallic (non-magnetic) interface, $\theta_T \approx 0$ independent of θ_I
 \Rightarrow exponentially decaying wave normal to interface.

Derivation of Fresnel's eqns. with \underline{E} polarised normal and parallel to plane of incidence [makes use of $H = nE/(\mu c)$]. Simplified form in the case of non-magnetic media. Dependence of phase on relative values of refractive indices. Brewster angle. Normal incidence. Reflection (R) and transmission (T) coefficients. $R + T = 1$.

Low pressure ionised gases - plasmas

Examples of plasmas. Response of low pressure (\Rightarrow no collisions/energy transfer) plasma to a plane EM field. Plasma conductivity = $-iNq^2/(\omega m)$. No net absorption of energy once oscillation established due to relative phase of E, J . Use of $k^2 = \omega^2 \mu \epsilon - i\mu \sigma \omega$ to obtain $k^2 = (1 - \omega_p^2/\omega^2)\omega^2/c^2$ with $\omega_p = (Nq^2/m\epsilon_o)^{1/2}$ - the plasma frequency. In hertz, $\nu_p \approx 9N^{1/2}$ Hz. Propagating ($\omega > \omega_p$) and attenuated ($\omega < \omega_p$) solution regimes. Decay length.

Results for the phase and group velocities and refractive index are:

Phase velocity, $v_{\text{ph}} = \frac{c}{\left(1 - \frac{\omega_p^2}{\omega^2}\right)^{\frac{1}{2}}}$

Group velocity, $v_g = c \left(1 - \frac{\omega_p^2}{\omega^2}\right)^{\frac{1}{2}}$

Refractive index, $n = \left(1 - \frac{\omega_p^2}{\omega^2}\right)^{\frac{1}{2}}$

(Above applying when $\omega > \omega_p$.)

Ionosphere used as a particular example. Handout giving review of implications for terrestrial (D, E, F, F₁, F₂ layers) and satellite communications.

Radiation theory

Solution of Maxwell's eqns. in time-independent case. Vector potential, \underline{A} . Choice of Coulomb gauge in order to obtain expression for \underline{A} by analogy with more familiar expression for scalar potential, V :

$$V(r_2) = \frac{1}{4\pi\epsilon_0} \int \frac{\rho(r_1)d\tau_1}{|r_2 - r_1|}$$

$$\underline{A}(r_2) = \frac{\mu_0}{4\pi} \int \frac{\underline{J}(r_1)d\tau_1}{|r_2 - r_1|}$$

Generalisation to time-dependent case - "common sense" approach used rather than deep theoretical study. $\underline{B} = \nabla \times \underline{A}$, $\underline{E} = -\nabla V - \frac{\partial \underline{A}}{\partial t}$.

Hertzian/oscillating electric dipole (magnetic dipole not considered). Expression for \underline{A} obtained. Expression for $\underline{B} = \nabla \times \underline{A}$ obtained via derivation on handout. Small r (Biot-Savart-like) and large r (radiation term) limits discussed. \underline{E} , \underline{B} and \underline{S} patterns discussed using spherical co-ordinate system expressions. Radiation resistance (R_r), directivity (D) and beam width (W) defined using Hertzian dipole as an example ($R_r = 197 \Omega$, $D = 3/2$, $W = 90^\circ$). Realistic antenna/aerial radiating systems, e.g. half-wave dipole ($R_r = 73 \Omega$, $D = 1.64$, $W = 78^\circ$) including expressions for \underline{E} , \underline{B} - not derived, however, $\lambda/4$ monopole + earth screen. Broadside and end-fire arrays briefly discussed and simple reflector + dipole + directors (parasitic element) type Yagi-Uda array as used for tv reception.